### 3W Class-D Stereo Audio Amplifier

## **General Description**

The HM8403 is a 3W, class-D audio amplifier. It offers low THD+N, allowing it to achieve high-quality sound reproduction. The new filterless rchitecture allows the device to drive the speaker directly, requiring no low-pass output filters, thus to save the system cost and PCB area. With the same numbers of external components, the efficiency of the HM8403 is much better than that of class-AB cousins. It can extend the battery life, ideal for portable applications.

## Features

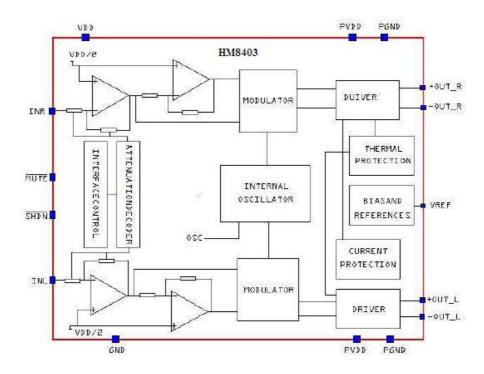
- 3W output at 10% THD with a 4ΩLoad and 5V Power Supply
- Filterless, Low Quiescent Current and Low EMI
- Low THD+N
- Superior Low Noise
- Efficiency up to 90%
- Short Circuit Protection
- Thermal Shutdown
- Few External Components to Save the Space and Cost
- Pb-Free

## **Physical Map**



## Applications

- LCD Monitors 1 TV Projectors
- Notebook Computers
- Portable Speakers
- Portable DVD Players, Game Machines
- Cellular Phones1Speaker Phones



## **Block Diagram**

# **Ordering Information**

Part Number	Package Type	Shipping Type	Note
HM8403S	SOP16	50 Units/Tube	
HM8403Q	QFN4X4	3000 Units/Reel	
HM8403QF	QFN3X3	3000 Units/Reel	

# **Typical Application**

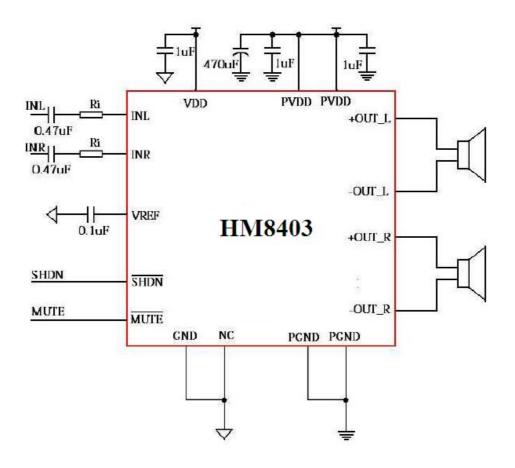


FIGURE 1. Typical Audio Amplifier Application Circuit

### **Connetction Diagrams**

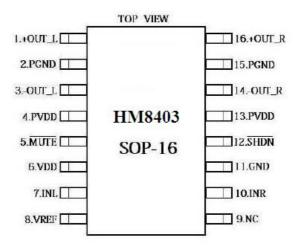


FIGURE2. SOP-16 Package

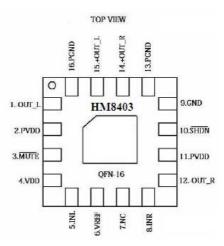


FIGURE3. QFN-16 Package

# **Terminal Functions**

Pin I	Number	Pin Name	Discription	
15	1	+out_L	Left Channel Positive output	
16	2	PGND	Power GND	
1	3	-out_L	Left Channel Negative output	
2	4	PVDD	Power VDD	
3	5	MUTE	Mute Control Input (active low)	
4	6	VDD	Analog VDD	
5	7	INL	Left Channel Input	
6	8	VREF	Internal analog reference, connect a bypass capacitor from VREF to GND	
7	9	NC	No connect	
8	10	INR	Right Channel Input	
9	11	GND	Analog GND	
10	12	SHDN	Shutdown Control Input (active low)	
11	13	PVDD	Power VDD	
12	14	-out_R	Right Channel Negative output	
13	15	PGND	Power GND	
14	16	+out_R	Right Channel Positive output	

## **Absolute Maximum Ratings**

These are stress ratings only and functional operation is not implied Exposure to absolute maximum ratings for prolonged time periods may affect device reliability All voltages are with respect to ground.

Supply Voltage	6.0V
Input Voltage	-0.3V to VDD+0.3V
Operation Temperature Range	<b>-40</b> ℃ to 85℃
Maximum Junction Temperature	<b>150</b> ℃
Operatio Junction Temperature	<b>-40℃ to 125℃</b>
Storage Temperture	<b>-65</b> ℃ to 150℃
Soldering temperature	<b>300</b> ℃, <b>5sec</b>

## **Recommended Operating Conditions**

Supply voltage Range	2.5V to 5.5V
Operation Temperature Range	40 to 85
Junction Temperature Range	40 to 125

## **Thermal Information**

Parameter	Symbol	Package	Maximum	Unit
Thermal Resis tance (Junction to Ambient)	θJ	SoP-16	110	°C1
Thermal Resis tance (Junction to Case)	θJ	SoP-16	23	°C1

**Electrical Characteristics** 

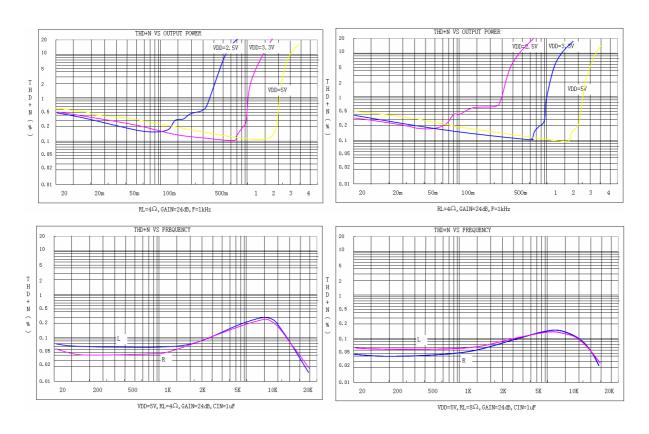
 $V_{\text{DD}}$ =5V, Gain=24dB, R<sub>L</sub>=80, T<sub>A</sub>=25°C, unless otherwise noted.

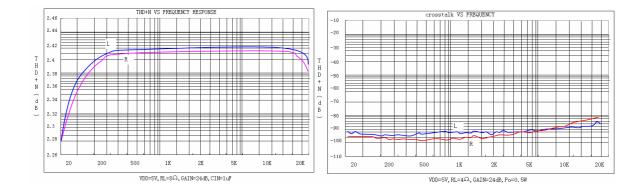
88							
Sym	Parameter	Test Conditions	MI	TYP	М	U	

VIN	Supply Power			2.5		5.5	V
Po	output Power		VDD=5.0V		3.2		
		THD+N=10%,f=1kHz, RL=4	VDD=3.6V		1.6		w
			VDD=3.		1.3		••
			VDD=5.		2.5		
			VDD=3.6V		1.3		
		THD+N=1%,f=1kHz,	VDD=3.		0.85		W
			VDD=5.		1.8		
			VDD=3.6V		0.9		
		THD+N=10%,f=1kHz,	VDD=3.		0.6		W
			VDD=5.		1.4		
			VDD=3.6V		0.72		
		THD+N=1%,f=1kHz,	VDD=3.		0.45		W
		VDD=5.0V,Po=0.5W,RL			0.15		
		VDD=3.6V,Po=0.5W,RL=8	f=1kHz		0.11		%
	Total Harmonic Distortion	VDD=5.0V,Po=1W,RL=			0.15		
THD	Plus	VDD=3.6V,Po=1W,RL=4 Ω	f=1kHz		0.11		%
Gv	Gain				24		dB
		VDD=5.0V, Inputs	f=100H		-59		
PSR	Power Supply Ripple	ac-grounded with	f=1kHz		-58		d
Cs	Crosstalk	VDD=5V,Po=0.5W,RL=8	F=1kHz		-95		dB
SNR	Signal-to-noise ratio	VDD=5V,	f=1kHz		80		dB
		VDD=5V, Inputs	A-weighti		100		
V	output noise	ac-grounded with	No		150		μ
Dyn	Dynamic range	VDD=5.0V, THD=1%	f=1kHz		90		dB
		RL=8 Ω , THD=10%			8		
η	Efficie	RL=4 Ω , THD=10%	f=1kHz		83		%
		VDD			16		
		VDD=3.6V			10		
Ι	Quiescent Current	VDD	No load		8		m
Sym	Parame	Test		MI	ΤY	М	UNI

IMUT	Muting Current	VDD=5.0V	VMUTE		3.5		mA
ISD	Shutdown Current	VDD=2.5V to 5.5V	Vsd=0.3V		<1		μA
	Static Drain-to-source		PMoS		18		
Rds	on-state	IDS =500mA,Vgs=5V	NMoS		140		mo
fsw	Switching Frequency	VDD=3V to 5V			260		kHz
Vos	output offset Voltage	Vin=0V, VDD=5V			10		mV
VIH	Enable Input High Voltage	VDD=5.0V		1.5	1.4		
VIL	Enable Input Low	VDD=5.0V			0.7	0.4	V
VIH	MUTE Input High Voltage	VDD=5.0V		1.5	1.4		
VIL	MUTE Input Low Voltage	VDD=5.0V			0.7	0.4	V
oTP	over Temperature				140		
oTH	over Temperature	No Load, Junction	VDD=5V		30		°C

# **Operating Characteristics**





## **Application Notes**

1. When the HM8403 works with LC filters, it should be connected with the speaker before it's powered on, otherwise it will be damaged easily.

2. When the HM8403 works without LC filters, it's better to add a ferrite chip bead at the outgoing line of speaker for suppressing the possible electromagnetic interference.

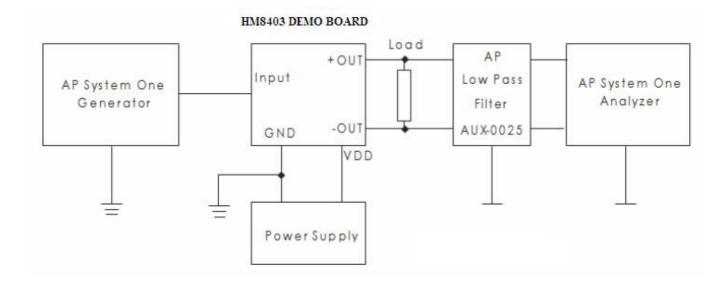
3. The recommended operating voltage is 5.5V. When the HM8403 is powered with 4 battery cells,

it should be noted that the voltage of 4 new dry or alkaline batteries is over 6.0V, higher than its operation voltage, which will probably damage the device. Therefore, it's recommended to use either 4 Ni-MH(Nickel Metal Hydride) rechargeable batteries or 3 dry or alkaline batteries.

4. one should not make the input signal too large. Large signal can cause the clipping of output signal when increasing the volume. This will damage the device because of big gain of the HM8403.

5. When testing the HM8403without LC filters by using resistor instead of speaker as the output load, the test results, e.g. THD or efficiency, will be worse than those of using speaker as load.

## **Test Setup for Performance Testing**



#### FIGURE 6. Test Set-Up For Graphs

#### Notes

1. The AP AUX-0025 low pass filter is necessary for class-D amplifier measurement with AP analyzer.

2. Two 22 $\mu$ H inductors are used in series with load resistor to emulate the small speaker for efficiency

measurement.

## **Application Information**

### **Maximum Gain**

As shown in block diagram (page 2), the HM8403 has two internal amplifier stages. The first stage's gain is externally configurable, while the second stage's is internally fixed. The closed-loop gain of the first stage is set by selecting the ratio of R to R while the second stage's gain is fixed at 2x. The output of amplifier 1 serves as the input to amplifier 2, thus the two a m p l i f i e r s produce si g n a l s i d e n t i c a l i n magnitude, but different in phase by 180°. Consequently, the differential gain for the IC is

### A =20\*log [2\*(R /R )]

The HM8403 sets maximum R =142k , minimum R =18k , so the maximum closed-gain is 24dB.

### **Mute Operation**

The pin is an input for controlling the output state of the HM8403. A logic low on this pin disables the outputs, and a logic high on this pin enables the outputs. This pin may be used as a quick disable or enable of the outputs without a volume fade. Quiescent current is listed in the electrical characteristic table. The pin can be left floating due to the internal pull-up.

### Shutdown operation

In order to reduce power consumption while not in use, the HM8403 contains shutdown circuitry to turn off the amplifier's bias circuitry. This shutdown feature turns the amplifier off when logic low is applied to the pin. By switching the pin connected to GND, the HM8403 supply current draw will be minimized in idle mode. The pin can be left floating due to the internal pull-up.

## Power supply decoupling

The HM8403 is a high performance CMOS audio amplifier that requires adequate power supply decoupling to ensure the output THD and PSRR as low as possible. Power supply decoupling affects low frequency response. Optimum decoupling is achieved by using two capacitors of different types targeting to different types of noise on the power supply leads. For higher frequency transients, spikes, or digital hash on the line, a good low equivalent-seriesresistance (ESR) ceramic capacitor, typically  $1.0\mu$ F, works best, placing it as close as possible to the device V terminal. For filtering lowerfrequency noise signals, a large capacitor of  $20\mu$ F (ceramic) or greater is recommended, placing it near the audio power amplifier.

## Input Capacitor (Ci)

Large input capacitors are both expensive and space hungry for portable designs. Clearly, a certain sized capacitor is needed to couple in low frequencies without severe attenuation. But in many cases the speakers used in portable systems, whether internal or external, have little ability to reproduce signals below 100Hz to 150Hz. Thus, using a large input capacitor may not increase actual system performance. In this case, input capacitor (Ci ) and input resistance (R i) of the amplifier form a high-pass filter with the corner frequency determined by equation below,

$$f_c = \frac{1}{2\pi R_i C_i}$$

In addition to system cost and size, click and pop performance is affected by the size of the input coupling capacitor, C . A larger input coupling capacitor requires more charge to reach its quiescent DC voltage (nominally 1/2 V). This charge comes from the internal circuit via the feedback and is apt to create pops upon device enable. Thus, by minimizing the capacitor size based on necessary low frequency response, turn-on pops can be minimized.

### Analog Reference Bypass Capacitor (CBYP)

The Analog Reference Bypass Capacitor (CBYP) is the most critical capacitor and serves several important functions. During start-up or recovery from shutdown mode, C determines the rate at which the amplifier starts up. The second function is to reduce noise caused by the power supply coupling into the output drive signal. This noise is from the internal analog reference to the amplifier, which appears as degraded PSRR and THD+N.

A ceramic bypass capacitor (CBYP) with values of  $0.47\mu$ F to  $1.0\mu$ F is recommended for the best THD and noise performance. Increasing the bypass capacitor reduces clicking and popping noise from power on/off and entering and leaving shutdown.

## Under Voltage Lock-out (UVLO)

The HM8403 incorporates circuitry designed to detect low supply voltage. When the supply voltage drops to 2.0V or below, the HM8403 outputs are disabled, and the device comes out of this state and starts to normal function when Vdd>=2.2V.

## **Short Circuit Protection (SCP)**

The HM8403 has short circuit protection circuitry on the outputs to prevent damage to the device when output-to-output or output-to-GND short occurs. When a short circuit is detected on the outputs, the outputs are disabled immediately. If the short was removed, the device activates again.

### **Over Temperature Protection**

Thermal protection on the HM8403 prevents the device from damage when the internal die temperature exceeds 140°C. There is a 15 degree tolerance on this trip point from device to device. Once the die temperature exceeds the thermal set point, the device outputs are disabled. This is not a latched fault. The thermal fault is cleared

once the temperature of the die is reduced by 30°C. This large hysteresis will prevent motor boating sound well and the device begins normal operation at this point without external system intervention.

### How to Reduce EMI (Electro Magnetic Interference)

A simple solution is to put an additional capacitor 1000 F at power supply terminal for power line coupling if the traces from amplifier to speakers are short (<20cm).

Most applications require a ferrite bead filter as shown in Figure 2. The ferrite filter reduces EMI of around 1 MHz and higher. When selecting a ferrite bead, choose one with high impedance at high frequencies, and low impedance at low frequencies.

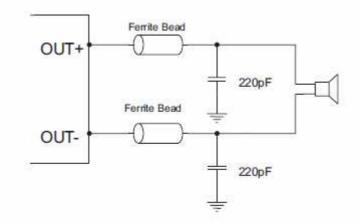
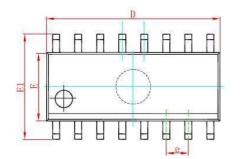
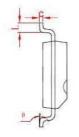


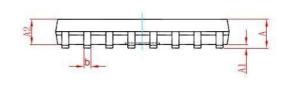
Figure 2: Ferrite Bead Filter to reduce EMI

# **Package Information**

1、 SOP16

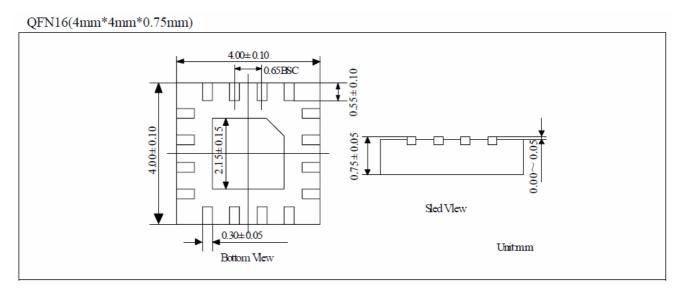




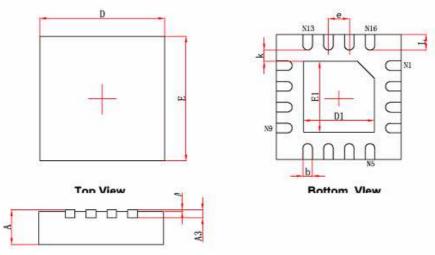


800 N N	Dimensions I	n Millimeters	Dimension	s In Inches
Symbol	Min	Max	Min	Max
A	1.350	1.750	0.053	0.069
A1	0.100	0. 250	0.004	0.010
A2	1.350	1.550	0.053	0.061
b	0.330	0.510	0.013	0.020
с	0. 170	0.250	0.007	0.010
D	9.800	10. 200	0.386	0.402
E	3.800	4.000	0.150	0.157
E1	5.800	6.200	0. 228	0.244
е	1.27	0 (BSC)	0.05	0 (BSC)
L	0.400	1.270	0.016	0.050
θ	0°	8°	0°	8°

### 2.QFN4X4



3.QFN3X3



Side	View
Juc	410.44

Combal	Dimensions	n Millimeters	Dimension	s In Inches
Symbol	Min.	Max.	Min.	Max.
A	0.700/0.800	0.800/0.900	0.028/0.031	0.031/0.035
A1	0.000	0.050	0.000	0.002
A3	0.203	REF.	0.008	REF.
D	2.900	3.100	0.114	0.122
E	2.900	3.100	0.114	0.122
D1	1.600	1.800	0.063	0.071
E1	1.600	1.800	0.063	0.071
k	0.200MIN.		0.00	BMIN.
b	0.180	0.300	0.007	0.012
е	0.500	TYP.	0.020	TYP.
L	0.300	0.500	0.012	0.020